



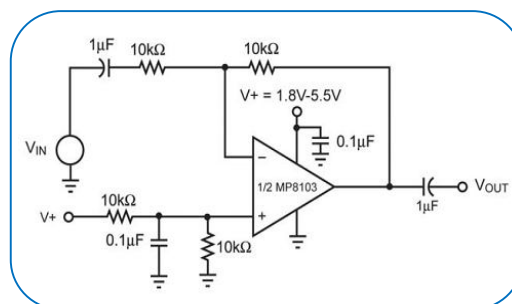
## STUDIES ON LOW-VOLTAGE ULTRA-LOW-POWER NETWORK

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### ABSTRACT :

In this paper, it's argued that there area unit smart reasons to decide on current because the information-carrying amount within the case of low-tension ultra-low-power style constraints. This paper focuses on the influence of transfer quality thereon selection. to get power-efficient transfer quality, indirect feedback is shown to be a decent different to ancient feedback techniques. 2 recently developed analog circuit techniques that each operate within the current domain and use indirect feedback area unit delineated , being the continuous-time dynamic-translinear technique and also the discrete-time switched-MOSFET technique



**KEYWORDS :** ancient feedback techniques , dynamic-translinear technique.

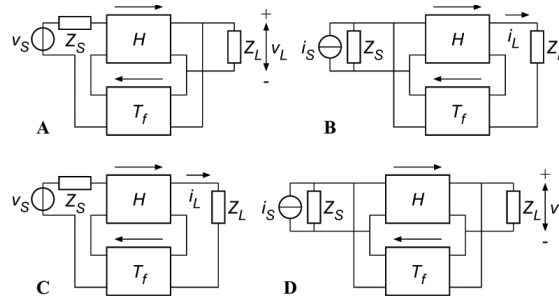
### INTRODUCTION :

Low-voltage circuit techniques area unit applied within the space of battery-operated systems. For movableness reasons, the scale of the instrumentation should be tiny, that necessitates the most integration of the signal process electronic equipment. However, because the size of batteries is currently changing into the limiting issue, the reduction of the facility dissipation has become an additional style constraint. As a consequence, the key purpose is to develop, at the same time, each low-tension (i.e. 1 – 1.5 V) and low-current (i.e. < one mA) operative integrated circuits so as to scale back the battery size. Another style criterion that has to be consummated is transfer quality. This quality is influenced by 2 totally different sorts of errors: random ones and systematic ones. By random errors we have a tendency to mean inaccuracies within the input-output relation caused by noise or interference. It is not possible to eliminate, their influence may be reduced by a correct style strategy. Systematic errors arise from network imperfections, like offset, non-linearity, quality of the device parameters, drift and temperature dependence. in all probability the foremost effective technique to scale back their influence, and therefore to get an correct transfer perform, is by means that of applying feedback, that permits United States of America to exchange the big gain provided by the (highly non-linear) active devices for quality provided by (usually linear) passive devices.[1-2] Design methods for the reduction of random errors and systematic errors area unit ordinarily not per style methods that take into consideration power dissipation, voltage vary and current vary. Therefore, it's the mix of transfer quality, low voltage and low power that has to be thought of throughout the entire style method.

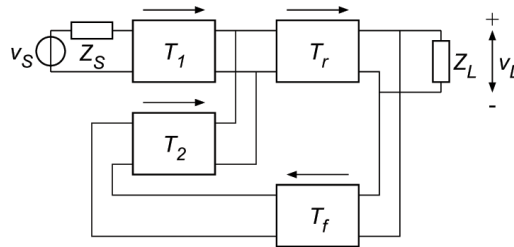
### INDIRECT FEEDBACK

As mentioned on top of, systematic errors may be reduced by means that of feedback. Fig. one shows the four basic ways in which of applying (single-loop) direct feedback by means that of 2 two-ports. If

all the transfer parameters of two-port H approach eternity, i.e., H may be a nullor, the signal ( $v_L$  or  $i_L$ ) is expounded to the signal ( $v_S$  or  $i_S$ ) because the inverse transfer perform of the feedback network  $T_f$ .



**Fig. 1. Four basic direct negative-feedback amplifiers: a voltage amplifier (A), a current amplifier (B), a transconductance amplifier (C) and a transimpedance amplifier (D).**



**Fig. 2. A voltage amplifier with negative feedback and indirect voltage comparison.**

In low-tension circuits, however, thanks to the restricted voltage swing, it's typically impracticable, or a minimum of not preferred, to attach 2 ports of those two-port networks asynchronous, therefore to sense the output current or to match the input voltage of a circuit directly. This happens in configurations A (at the input), B (at the output) and C (at each input and output). Hence, all direct-feedback configurations, except the transimpedance electronic equipment (configuration D), square measure less fitted to low-tension applications. To realize voltage, current and transconductance amplifiers, a helpful different to direct feedback is also a method referred to as indirect feedback [1]. In associate indirect-negative-feedback circuit, the output and/or the input stage is derived, in order that it's the same input-output relation, and therefore the feedback signal is taken from and/or fed back thereto copy. Thus, it's attainable to get a circuit response that is decided by the feedback network solely, presumptuous that the repeating doesn't introduce errors. Voltage electronic equipment, a current electronic equipment and a transconductance electronic equipment, all victimization the indirect negative-feedback principle, square measure delineated in Figures a pair of, 3 and 4. It may be seen that series-connected ports square measure currently avoided all told configurations.[4]

Again, if all the transfer parameters of two-port  $T_r$  approach infinity,  $T_2 = T_1$  and  $T_4 = T_3$ , the output signal ( $v_L$  or  $i_L$ ) is related to the input signal ( $v_S$  or  $i_S$ ) as the inverse transfer function of the feedback network  $T_f$ .

**PROCESSING IN THE CURRENT DOMAIN**

We currently investigate however applying indirect feedback relates to the selection of the electrical quantities within the system. In electronic circuits, indirect voltage comparison leads to a doubled power density spectrum of the equivalent noise voltage at the input, as a result of the outputs of the direct and indirect input stages ar connected in parallel.[2,3] Indirect current sensing leads to a doubled power density spectrum of the noise current at the output, as a result of the direct and indirect output are placed in

parallel. In observe, typically the noise is most crucial at the input, therefore on it ground there could also be a preference for current sensing and so for current because the information-carrying amount.

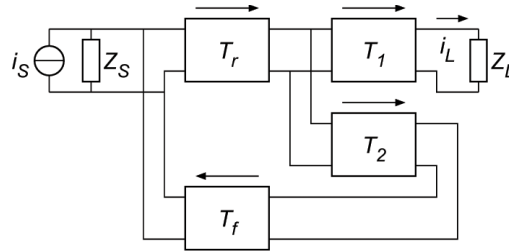


Fig. 3. A current amplifier with negative feedback and indirect current sensing.

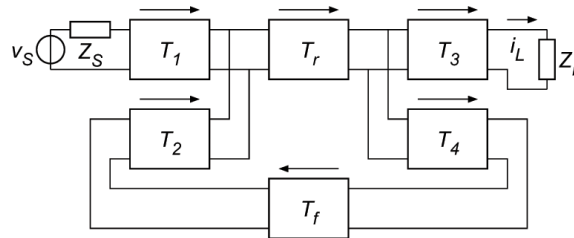


Fig. 4. A transconductance amplifier with negative feedback

## DISCUSSION

Another disadvantage of the use of voltage as the information-carrying quantity is that, when the circuits are “voltage-driven,” i.e., from a low-impedance source, the equivalent input noise voltage is predominantly the result of the input noise voltage of both input stages. For bipolar transistors and CMOS transistors in weak inversion,[5] this input noise voltage is inversely proportional to the bias (collector or drain) current, and thus, in order to obtain a low input noise voltage, these bias currents must be rather large. This, of course, is in sharp contrast with our low-power requirement.

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